TEXAS INSTRUMENTS

LM1085

3A Low Dropout Positive Regulators

General Description

The LM1085 is a series of low dropout positive voltage regulators with a maximum dropout of 1.5V at 3A of load current. It has the same pin-out as National Semiconductor's industry standard LM317.

The LM1085 is available in an adjustable version, which can set the output voltage with only two external resistors. It is also available in three fixed voltages: 3.3V, 5.0V and 12.0V. The fixed versions integrate the adjust resistors.

The LM1085 circuit includes a zener trimmed bandgap reference, current limiting and thermal shutdown.

The LM1085 series is available in TO-220 and TO-263 packages. Refer to the LM1084 for the 5A version, and the LM1086 for the 1.5A version.

Features

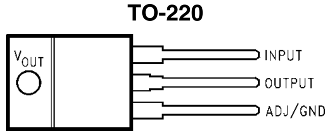
- Available in 3.3V, 5.0V, 12V and Adjustable Versions
- Current Limiting and Thermal Protection
- Output Current 3A
- Line Regulation 0.015% (typical)
- Load Regulation 0.1% (typical)

Applications

- High Efficiency Linear Regulators
- Battery Charger
- Post Regulation for Switching Supplies
- Constant Current Regulator
- Microprocessor Supply

Connection Diagrams

TO-220



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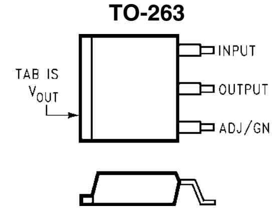
Top View

TO-263



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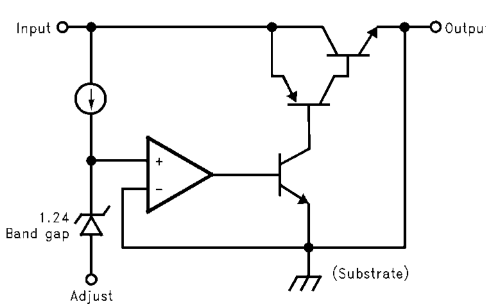
Top View

Basic Functional Diagram, Adjustable Version



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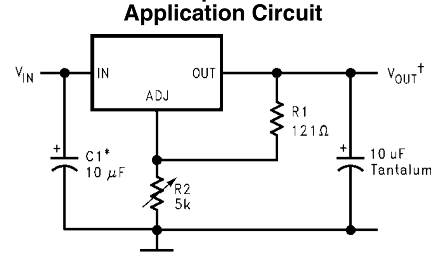
Application Circuit



*NEEDED IF DEVICE IS FAR FROM FILTER CAPACITORS

$^{\dagger}V_{OUT} = 1.25V\left(1 + \frac{R2}{R1}\right)$

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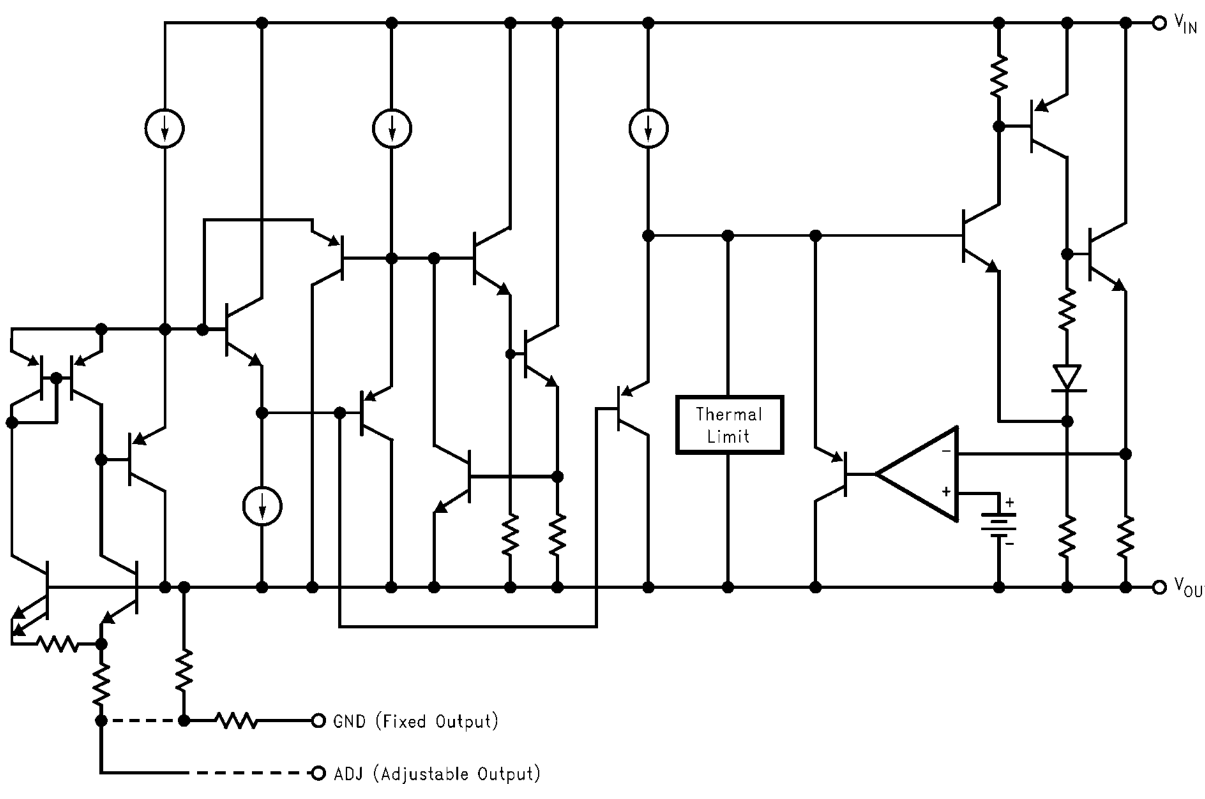
1.2V to 15V Adjustable Regulator

Ordering Information

Package	NSC Drawing	Temperature Range	Output Voltage	Part Number	Transport Media
TO-263 3-lead	TS3B	−40°C to +125°C	ADJ	LM1085IS-ADJ	Rail of 45
				LM1085ISX-ADJ	Reel of 500
			12	LM1085IS-12	Rail of 45
				LM1085ISX-12	Reel of 500
			3.3	LM1085IS-3.3	Rail of 45
				LM1085ISX-3.3	Reel of 500
			5.0	LM1085IS-5.0	Rail of 45
				LM1085ISX-5.0	Reel of 500
TO-220 3-lead	T03B	−40°C to + 125°C	ADJ	LM1085IT-ADJ	Rail of 45
			12	LM1085IT-12	Rail of 45
			3.3	LM1085IT-3.3	Rail of 45
			5.0	LM1085IT-5.0	Rail of 45

Simplified Schematic



Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.

Maximum Input to Output Voltage Differential	
LM1085-ADJ	29V
LM1085-12	18V
LM1085-3.3	27V
LM1085-5.0	25V
Power Dissipation (Note 2)	Internally Limited
Junction Temperature (T_J)(Note 3)	150°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature	260°C, to 10 sec
ESD Tolerance (Note 4)	2000V

Operating Ratings (Note 1)

Junction Temperature (T_J)(Note 3)	−40°C to 125°C

Electrical Characteristics

Limits in standard type are for T_J = 25°C only; limits in **boldface type** apply over the operating junction temperature (T_J) range of **-40°C to +125°C**. Minimum and Maximum limits are guaranteed through test, design, or statistical correlation. Typical values represent the most likely parametric norm at T_J = 25°C, and are provided for reference purposes only.

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
V_{REF}	Reference Voltage (Note 7)	LM1085-ADJ I_{OUT} = 10mA, $V_{IN}-V_{OUT}$ = 3V 10mA ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$, 1.5V ≤ ($V_{IN}-V_{OUT}$) ≤ 15V	1.238 **1.225**	1.250 **1.250**	1.262 **1.270**	V
V_{OUT}	Output Voltage (Note 7)	LM1085-3.3 I_{OUT} = 0mA, V_{IN} = 5V 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$, 4.8V ≤ V_{IN} ≤ 15V	3.270 **3.235**	3.300 **3.300**	3.330 **3.365**	V
		LM1085-5.0 I_{OUT} = 0mA, V_{IN} = 8V 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$, 6.5V ≤ V_{IN} ≤ 20V	4.950 **4.900**	5.000 **5.000**	5.050 **5.100**	V
		LM1085-12 I_{OUT} = 0mA, V_{IN} = 15V 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$, 13.5V ≤ V_{IN} ≤ 25V	11.880 **11.760**	12.000 **12.000**	12.120 **12.240**	V
ΔV_{OUT}	Line Regulation (Note 8)	LM1085-ADJ, I_{OUT} =10mA, 1.5V ≤ (V_{IN}-V_{OUT}) ≤ 15V		0.015 **0.035**	0.2 **0.2**	%
		LM1085-3.3, I_{OUT} = 0mA, 4.8V ≤ V_{IN} ≤ 15V		0.5 **1.0**	6 **6**	mV
		LM1085-5.0, I_{OUT} = 0mA, 6.5V ≤ V_{IN} ≤ 20V		0.5 **1.0**	10 **10**	mV
		LM1085-12, I_{OUT} =0mA, 13.5V ≤ V_{IN} ≤ 25V		1.0 **2.0**	25 **25**	mV
ΔV_{OUT}	Load Regulation (Note 8)	LM1085-ADJ, (V_{IN}-V_{OUT}) = 3V, 10mA ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$		0.1 **0.2**	0.3 **0.4**	%
		LM1085-3.3, V_{IN} = 5V, 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$		3 **7**	15 **20**	mV
		LM1085-5.0, V_{IN} = 8V, 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$		5 **10**	20 **35**	mV
		LM1085-12, V_{IN} = 15V, 0 ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$		12 **24**	36 **72**	mV
V_{DO}	Dropout Voltage (Note 9)	LM1085-ADJ, 3.3, 5, 12 ΔV_{REF}, ΔV_{OUT} = 1%, I_{OUT} = 3A		**1.3**	**1.5**	V

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
I_{LIMIT}	Current Limit	LM1085-ADJ, V_{IN}–V_{OUT} = 5V	**3.2**	**5.5**		A
		LM1085-ADJ, V_{IN}–V_{OUT} = 25V	**0.2**	**0.5**		
		LM1085-3.3, V_{IN} = 8.0V	**3.2**	**5.5**		A
		LM1085-5.0, V_{IN} = 10V	**3.2**	**5.5**		A
		LM1085-12, V_{IN} = 17V	**3.2**	**5.5**		A
	Minimum Load Current (Note 10)	LM1085-ADJ, V_{IN} –V_{OUT} = 25V		**5.0**	**10.0**	mA
I_{GND}	Quiescent Current	LM1085-3.3, V_{IN} ≤ 18V		**5.0**	**10.0**	mA
		LM1085-5.0, V_{IN} ≤ 20V		**5.0**	**10.0**	mA
		LM1085-12, V_{IN} ≤ 25V		**5.0**	**10.0**	mA
	Thermal Regulation	T_A = 25°C, 30ms Pulse		.004	0.02	%/W
	Ripple Rejection	f_{RIPPLE} = 120Hz, C_{OUT} = 25µF Tantalum, I_{OUT} = 3A				
		LM1085-ADJ C_{ADJ} = 25µF, (V_{IN}–V_O) = 3V	**60**	**75**		dB
		LM1085-3.3, V_{IN} = 6.3V	**60**	**72**		dB
		LM1085-5.0, V_{IN} = 8.0V	**60**	**68**		dB
		LM1085-12, V_{IN} = 15V	**54**	**60**		dB
I_{ADJ}	Adjust Pin Current	LM1085–ADJ		55	**120**	µA
ΔI_{ADJ}	Adjust Pin Current Change	LM1085–ADJ 10mA ≤ I_{OUT} ≤ $I_{FULL\ LOAD}$, 1.5V ≤ V_{IN}–V_{OUT} ≤ 25V		**0.2**	**5**	µA
	Temperature Stability			**0.5**		%
	Long Term Stability	T_A= 125°C, 1000 Hrs		0.3	1.0	%
	RMS Output Noise (% of V_{OUT})	10Hz ≤ f ≤ 10 kHz		0.003		%
θ_{JC}	Thermal Resistance (Junction-to-Case)	3-Lead TO-263	-	0.7	-	°C/W
		3-Lead TO-220	-	0.7	-	

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Power dissipation is kept in a safe range by current limiting circuitry. Refer to Overload Recovery in Application Notes.

Note 3: The maximum power dissipation is a function of $T_{J(max)}$, θ_{JA}, and T_A. The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(max)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly into a PC board. Refer to Thermal Considerations in the Application Notes.

Note 4: For testing purposes, ESD was applied using human body model, 1.5kΩ in series with 100pF.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

Note 7: $I_{FULL\ LOAD}$ is defined in the current limit curves. The $I_{FULL\ LOAD}$ Curve defines the current limit as a function of input-to-output voltage. Note that 30W power dissipation for the LM1085 is only achievable over a limited range of input-to-output voltage.

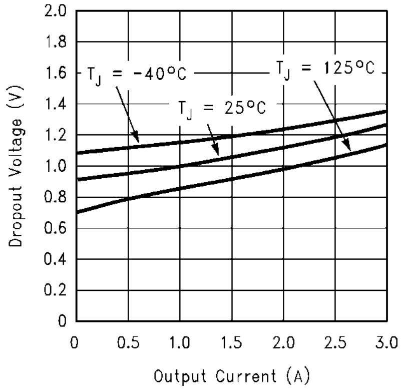
Note 8: Load and line regulation are measured at constant junction temperature, and are guaranteed up to the maximum power dissipation of 30W. Power dissipation is determined by the input/output differential and the output current. Guaranteed maximum power dissipation will not be available over the full input/output range.

Note 9: Dropout voltage is specified over the full output current range of the device.

Note 10: The minimum output current required to maintain regulation.

Typical Performance Characteristics

Dropout Voltage vs. Output Current



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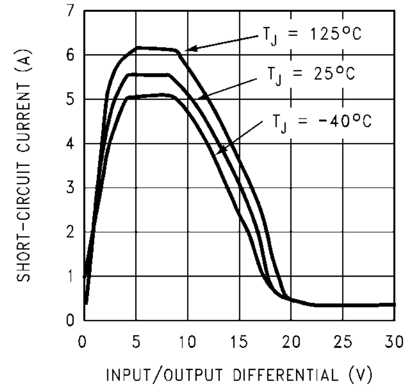
Short-Circuit Current vs. Input/Output Difference



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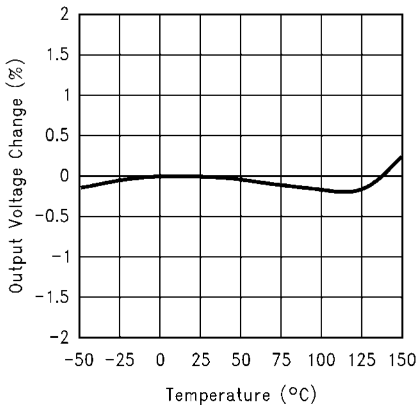
Percent Change in Output Voltage vs. Temperature



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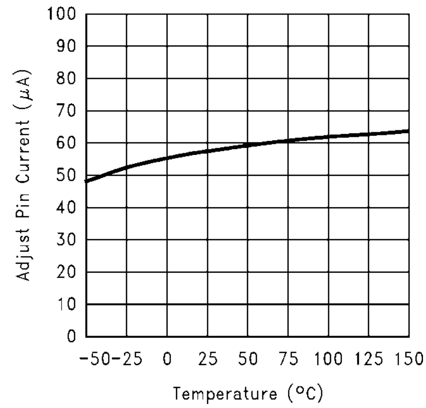
Adjust Pin Current vs. Temperature



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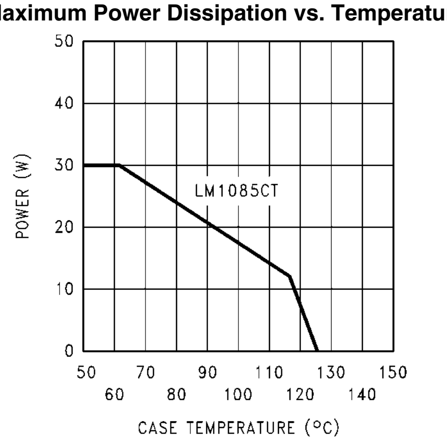
Maximum Power Dissipation vs. Temperature



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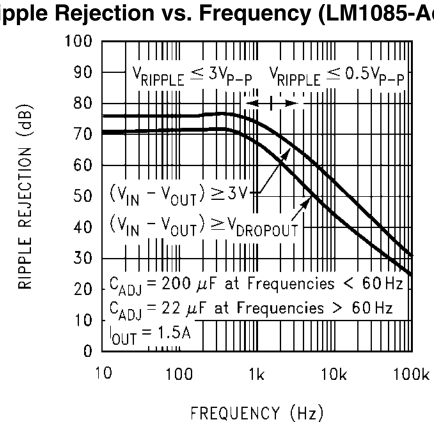
Ripple Rejection vs. Frequency (LM1085-Adj.)



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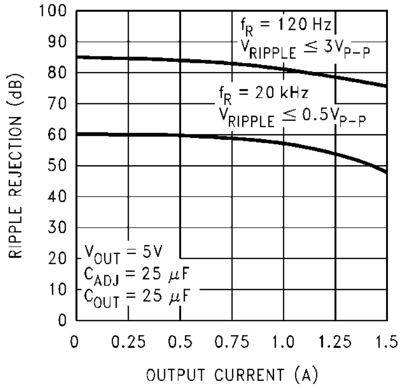
Ripple Rejection vs. Output Current (LM1085-Adj.)



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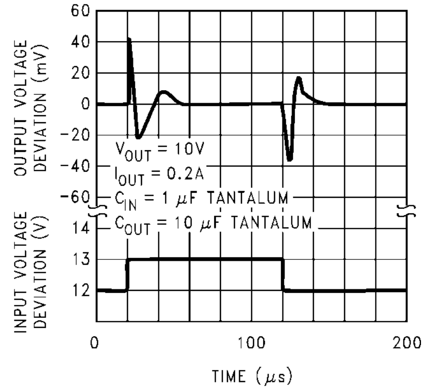
Line Transient Response



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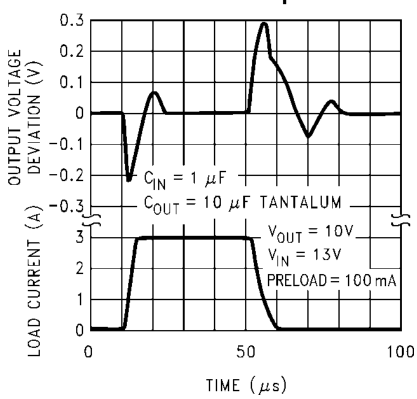
Load Transient Response



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Application Note

GENERAL

Figure 1 shows a basic functional diagram for the LM1085-Adj (excluding protection circuitry) . The topology is basically that of the LM317 except for the pass transistor. Instead of a Darlingtion NPN with its two diode voltage drop, the LM1085 uses a single NPN. This results in a lower dropout voltage. The structure of the pass transistor is also known as a quasi LDO. The advantage a quasi LDO over a PNP LDO is its inherently lower quiescent current. The LM1085 is guaranteed to provide a minimum dropout voltage 1.5V over temperature, at full load.

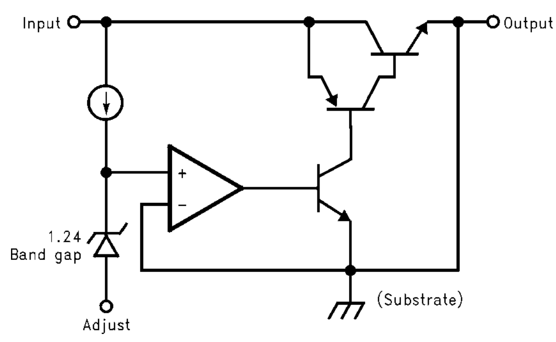


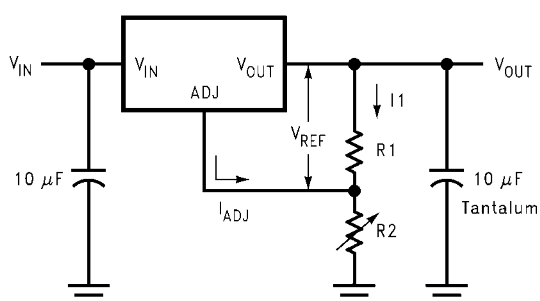
FIGURE 1. Basic Functional Diagram for the LM1085, excluding Protection circuitry

OUTPUT VOLTAGE

The LM1085 adjustable version develops at 1.25V reference voltage, (V_{REF}), between the output and the adjust terminal. As shown in figure 2, this voltage is applied across resistor R1 to generate a constant current I1. This constant current then flows through R2. The resulting voltage drop across R2 adds to the reference voltage to sets the desired output voltage.

The current I_{ADJ} from the adjustment terminal introduces an output error . But since it is small (120uA max), it becomes negligible when R1 is in the 100Ω range.

For fixed voltage devices, R1 and R2 are integrated inside the devices.



$$V_{OUT} = V_{REF}\left(1 + \frac{R2}{R1}\right) + I_{ADJ}\,R2$$

FIGURE 2. Basic Adjustable Regulator

STABILITY CONSIDERATION

Stability consideration primarily concern the phase response of the feedback loop. In order for stable operation, the loop must maintain negative feedback. The LM1085 requires a certain amount series resistance with capacitive loads. This series resistance introduces a zero within the loop to increase phase margin and thus increase stability. The equivalent series resistance (ESR) of solid tantalum or aluminum electrolytic capacitors is used to provide the appropriate zero (approximately 500 kHz).

The Aluminum electrolytic are less expensive than tantalums, but their ESR varies exponentially at cold temperatures; therefore requiring close examination when choosing the desired transient response over temperature. Tantalums are a convenient choice because their ESR varies less than 2:1 over temperature.

The recommended load/decoupling capacitance is a 10uF tantalum or a 50uF aluminum. These values will assure stability for the majority of applications.

The adjustable versions allows an additional capacitor to be used at the ADJ pin to increase ripple rejection. If this is done the output capacitor should be increased to 22uF for tantalums or to 150uF for aluminum.

Capacitors other than tantalum or aluminum can be used at the adjust pin and the input pin. A 10uF capacitor is a reasonable value at the input. See Ripple Rejection section regarding the value for the adjust pin capacitor.

It is desirable to have large output capacitance for applications that entail large changes in load current (microprocessors for example). The higher the capacitance, the larger the available charge per demand. It is also desirable to provide low ESR to reduce the change in output voltage:

$\Delta V = \Delta I \times ESR$

It is common practice to use several tantalum and ceramic capacitors in parallel to reduce this change in the output voltage by reducing the overall ESR.

Output capacitance can be increased indefinitely to improve transient response and stability.

RIPPLE REJECTION

Ripple rejection is a function of the open loop gain within the feed-back loop (refer to *Figure 1* and *Figure 2*). The LM1085 exhibits 75dB of ripple rejection (typ.). When adjusted for voltages higher than V_{REF}, the ripple rejection decreases as function of adjustment gain: (1+R1/R2) or V_O/V_{REF}. Therefore a 5V adjustment decreases ripple rejection by a factor of four (−12dB); Output ripple increases as adjustment voltage increases.

However, the adjustable version allows this degradation of ripple rejection to be compensated. The adjust terminal can be bypassed to ground with a capacitor (C_{ADJ}). The impedance of the C_{ADJ} should be equal to or less than R1 at the desired ripple frequency. This bypass capacitor prevents ripple from being amplified as the output voltage is increased.

$1/(2\pi * f_{RIPPLE} * C_{ADJ}) \leq R_1$

LOAD REGULATION

The LM1085 regulates the voltage that appears between its output and ground pins, or between its output and adjust pins. In some cases, line resistances can introduce errors to the voltage across the load. To obtain the best load regulation, a few precautions are needed.

Figure 3 shows a typical application using a fixed output regulator. Rt1 and Rt2 are the line resistances. V_{LOAD} is less than the V_{OUT} by the sum of the voltage drops along the line resistances. In this case, the load regulation seen at the R_{LOAD} would be degraded from the data sheet specification. To improve this, the load should be tied directly to the output terminal on the positive side and directly tied to the ground terminal on the negative side.

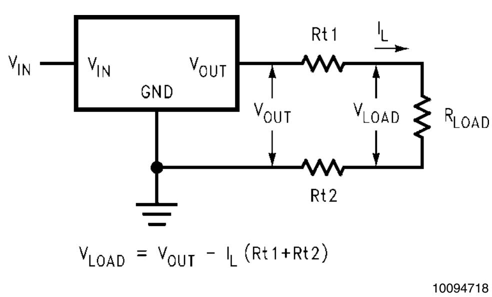


FIGURE 3. Typical Application using Fixed Output Regulator

When the adjustable regulator is used (*Figure 4*), the best performance is obtained with the positive side of the resistor R1 tied directly to the output terminal of the regulator rather than near the load. This eliminates line drops from appearing effectively in series with the reference and degrading regulation. For example, a 5V regulator with 0.05Ω resistance between the regulator and load will have a load regulation due to line resistance of 0.05Ω x I_L. If R1 (= 125Ω) is connected near the load the effective line resistance will be 0.05Ω (1 + R2/R1) or in this case, it is 4 times worse. In addition, the ground side of the resistor R2 can be returned near the ground of the load to provide remote ground sensing and improve load regulation.

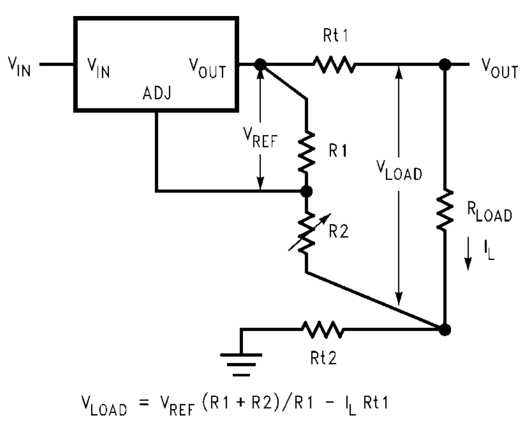


FIGURE 4. Best Load Regulation using Adjustable Output Regulator

PROTECTION DIODES

Under normal operation, the LM1085 regulator does not need any protection diode. With the adjustable device, the internal resistance between the adjustment and output terminals limits the current. No diode is needed to divert the current around the regulator even with a capacitor on the adjustment terminal. The adjust pin can take a transient signal of ±25V with respect to the output voltage without damaging the device.

When an output capacitor is connected to a regulator and the input is shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage of the regulator, and rate of decrease of V_{IN}. In the LM1085 regulator, the internal diode between the output and input pins can withstand microsecond surge currents of 10A to 20A. With an extremely large output capacitor (≥1000 µf), and with input instantaneously shorted to ground, the regulator could be damaged. In this case, an external diode is recommended between the output and input pins to protect the regulator, shown in *Figure 5*.

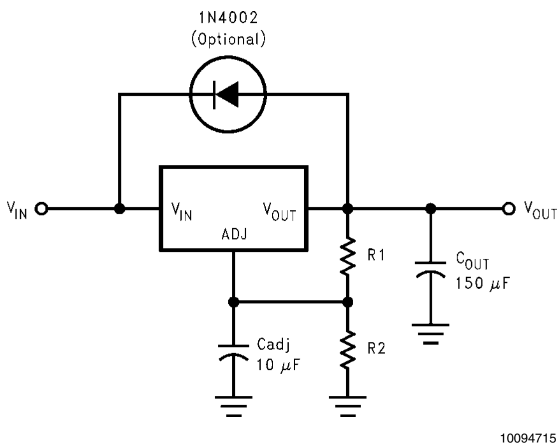


FIGURE 5. Regulator with Protection Diode

OVERLOAD RECOVERY

Overload recovery refers to regulator's ability to recover from a short circuited output. A key factor in the recovery process is the current limiting used to protect the output from drawing too much power. The current limiting circuit reduces the output current as the input to output differential increases. Refer to short circuit curve in the curve section.

During normal start-up, the input to output differential is small since the output follows the input. But, if the output is shorted, then the recovery involves a large input to output differential. Sometimes during this condition the current limiting circuit is slow in recovering. If the limited current is too low to develop a voltage at the output, the voltage will stabilize at a lower level. Under these conditions it may be necessary to recycle the power of the regulator in order to get the smaller differential voltage and thus adequate start up conditions. Refer to curve section for the short circuit current vs. input differential voltage.

THERMAL CONSIDERATIONS FOR THE TO-220 PACKAGE

ICs heats up when in operation, and power consumption is one factor in how hot it gets. The other factor is how well the heat is dissipated. Heat dissipation is predictable by knowing the thermal resistance between the IC and ambient (θ_{JA}). Thermal resistance has units of temperature per power (°C/W). The higher the thermal resistance, the hotter the IC.

The LM1085 specifies the thermal resistance for the TO-220 package as Junction to Case (θ_{JC}). In order to get the total resistance to ambient (θ_{JA}), two other thermal resistances must be added, one for case to heat-sink (θ_{CH}) and one for heatsink to ambient (θ_{HA}). The junction temperature can be predicted as follows:

$$T_J = T_A + (P_D \times (\theta_{JC} + \theta_{CH} + \theta_{HA}))$$

$$T_J = T_A + (P_D \times \theta_{JA})$$

where T_J is junction temperature, T_A is ambient temperature, and P_D is the power dissipation of the device.

Device power dissipation is calculated as follows:

$$P_D = \text{OUTPUT Section Dissipation} + \text{CONTROL Section Dissipation}$$

$$P_D = ((V_{IN} - V_{OUT}) \times I_{LOAD}) + ((V_{IN} - V_{OUT}) \times I_{GND})$$

Figure 6 shows the voltages and currents which are present in the circuit.

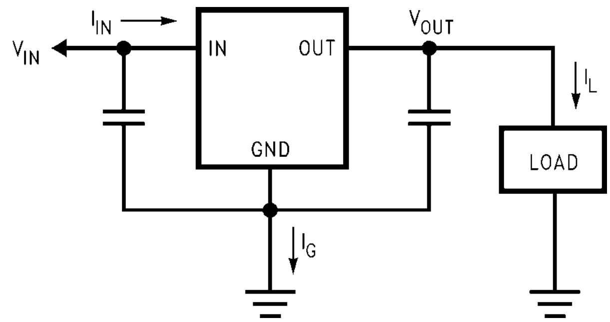


FIGURE 6. Power Dissipation Diagram

Once the devices power is determined, the maximum allowable ($\theta_{JA(max)}$) is calculated as:

$$\theta_{JA(MAX)} = T_{R(MAX)} / P_D$$

$$\theta_{JA(MAX)} = T_{J(MAX)} - T_{A(MAX)}) / P_D$$

The required heat sink is determined by calculating its required thermal resistance ($\theta_{HA(MAX)}$).

$$\theta_{HA(MAX)} = \theta_{JA(MAX)} - (\theta_{JC} + \theta_{CH})$$

If thermal compound is used, θ_{CH} can be estimated at 0.2 C/W. If the case is soldered to the heat sink, then a θ_{CH} can be estimated as 0 C/W.

If PC board copper is going to be used as a heat sink, then *Figure 7* can be used to determine the appropriate area (size) of copper foil required.

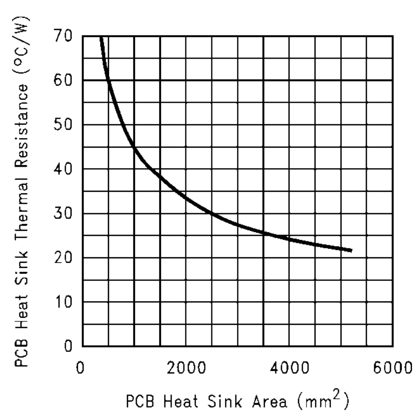


FIGURE 7. Heat sink thermal Resistance vs Area

THERMAL CONSIDERATION FOR THE TO-263 PACKAGE

Unlike the TO-220 package, the TO-263 package uses the printed circuit board as the heat sink to remove heat from the device. The device dissipation is:

$$P_D = \text{OUTPUT Section Dissipation} + \text{CONTROL Section Dissipation}$$

For the LM1085IS-x.x pre-set voltage versions, the dissipation can be calculated using:

$$P_D = ((V_{IN} - V_{OUT}) \times I_{LOAD}) + ((V_{IN} - V_{OUT}) \times I_{GND})$$

The LM1085IS-ADJ adjustable voltage version, the dissipation can be calculated using:

$$P_D = ((V_{IN} - V_{OUT}) \times I_{LOAD}) + ((V_{IN} - V_{OUT}) \times (V_{REF} / R1))$$

Current through the ADJ pin is sufficiently small such that any contribution to the device dissipation is so low that it can safely be ignored.

Maximum power dissipation of the LM1085IS depends on the total thermal resistance from the silicon junction through the package TAB (θ_{JC}), into the PC board, copper traces, and other materials, and then into the surrounding air (θ_{JA}), the maximum allowed operating junction temperature ($T_{J(MAX)}$) of 125°C, and the maximum ambient temperature ($T_{A(MAX)}$). The maximum power dissipation in the device is:

$$P_{D(MAX)} = (T_{J(MAX)} - T_{A(MAX)}) / (\theta_{JA}$$

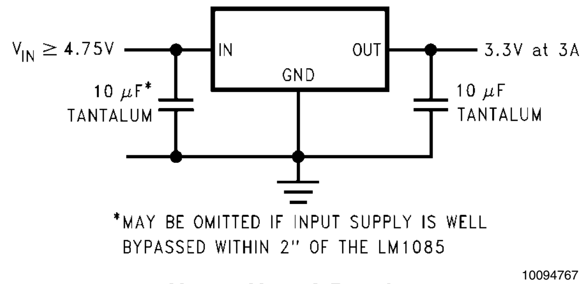
For the LM1085IS in the TO-263 3-pin package, the junction-to-case thermal rating, θ_{JC}, is 0.7°C/W, where the case is the bottom of the package at the center of the TAB. Typical junction-to-ambient thermal performance for the LM1085IS, using the JESD51 standards, is summarized in the following table:

BOARD TYPE	THERMAL VIAS	θ_{JA}
JEDEC 2-Layer (per JESD 51-3)	None	81 °C/W
JEDEC 4-Layer (per JESD 51-7)	0	59 °C/W
	2	31 °C/W
	4	27 °C/W
	8	24 °C/W
	12	23 °C/W

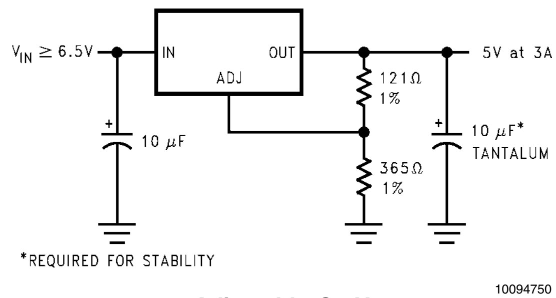
For more information refer to : “Application Note 1520 A Guide to Board Layout for Best Thermal Resistance for Exposed Packages”, TI Literature Number: SNVA183A

It is important to remember that the TAB of the LM1085IS package is internally conntected to device pin 2 (OUTPUT), so the copper area connected to the TAB must be isolated from all other potentials, including ground. The copper area connected to the TAB can be left floating, used as the primary V_{OUT} connection, or connected to device pin 2 (OUTPUT) .

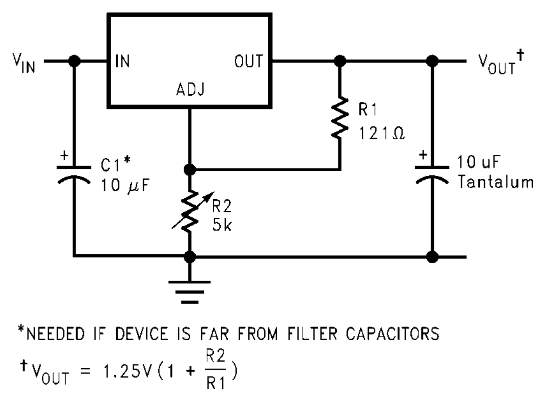
Typical Applications



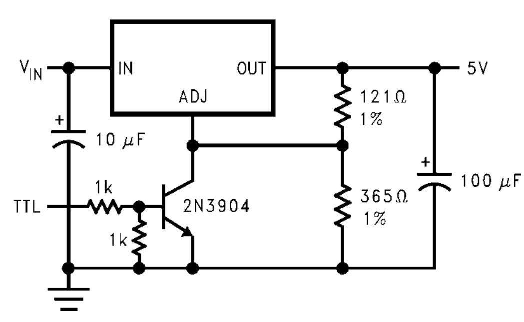
5V to 3.3V, 1.5A Regulator



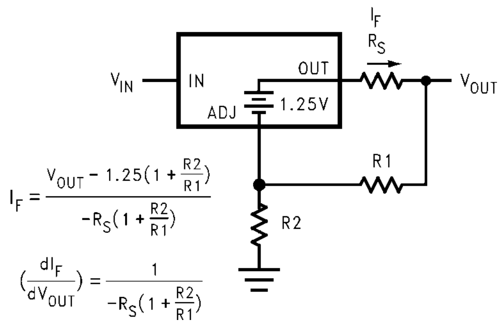
Adjustable @ 5V



1.2V to 15V Adjustable Regulator

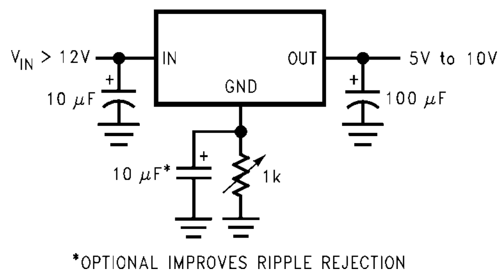


5V Regulator with Shutdown



10094754

Battery Charger



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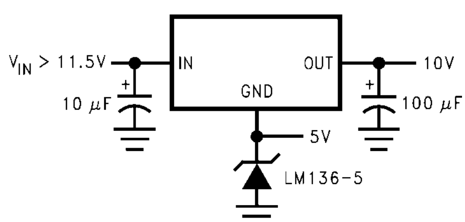
Adjustable Fixed Regulator



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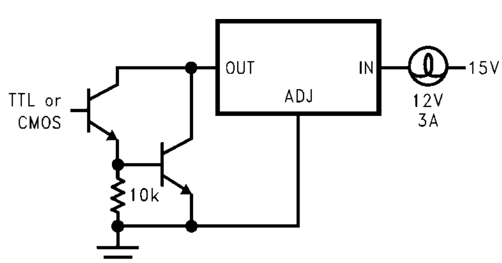
Regulator with Reference



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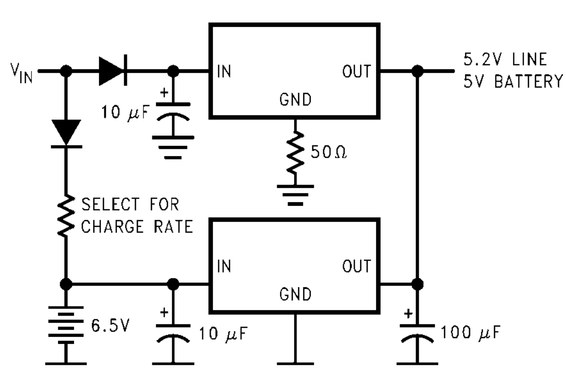
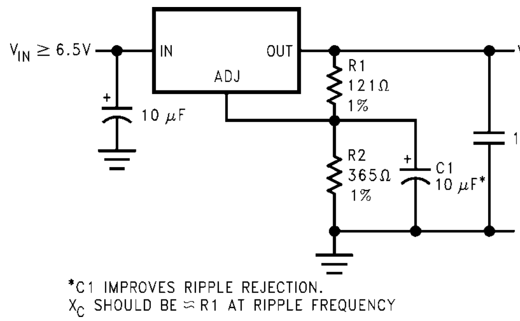
High Current Lamp Driver Protection



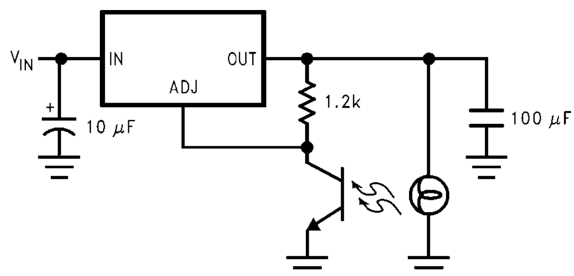
10094759

Battery Backup Regulated Supply



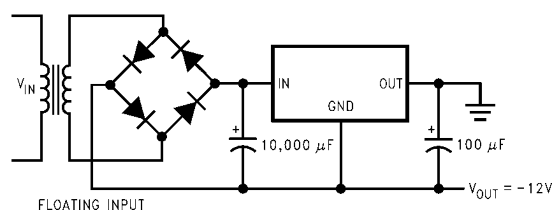
10094760

Ripple Rejection Enhancement



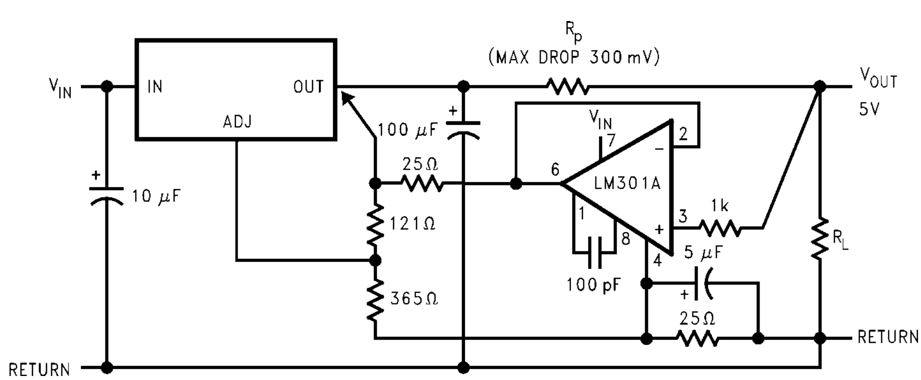
10094761

Automatic Light control



10094762

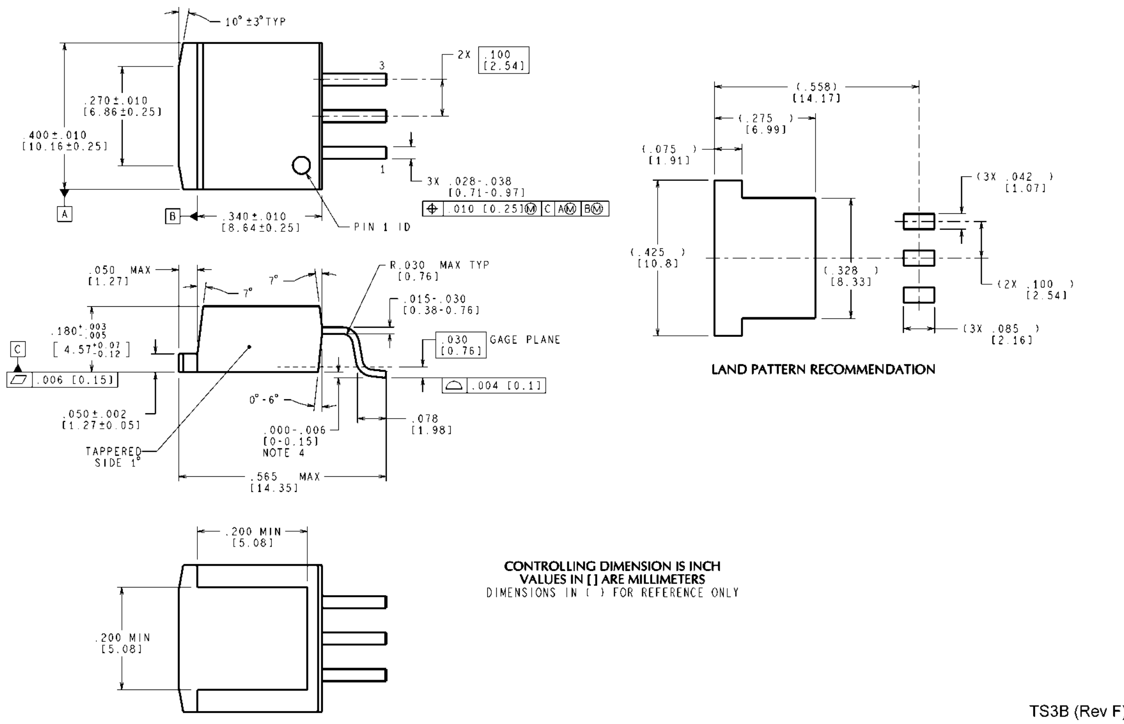
Generating Negative Supply voltage



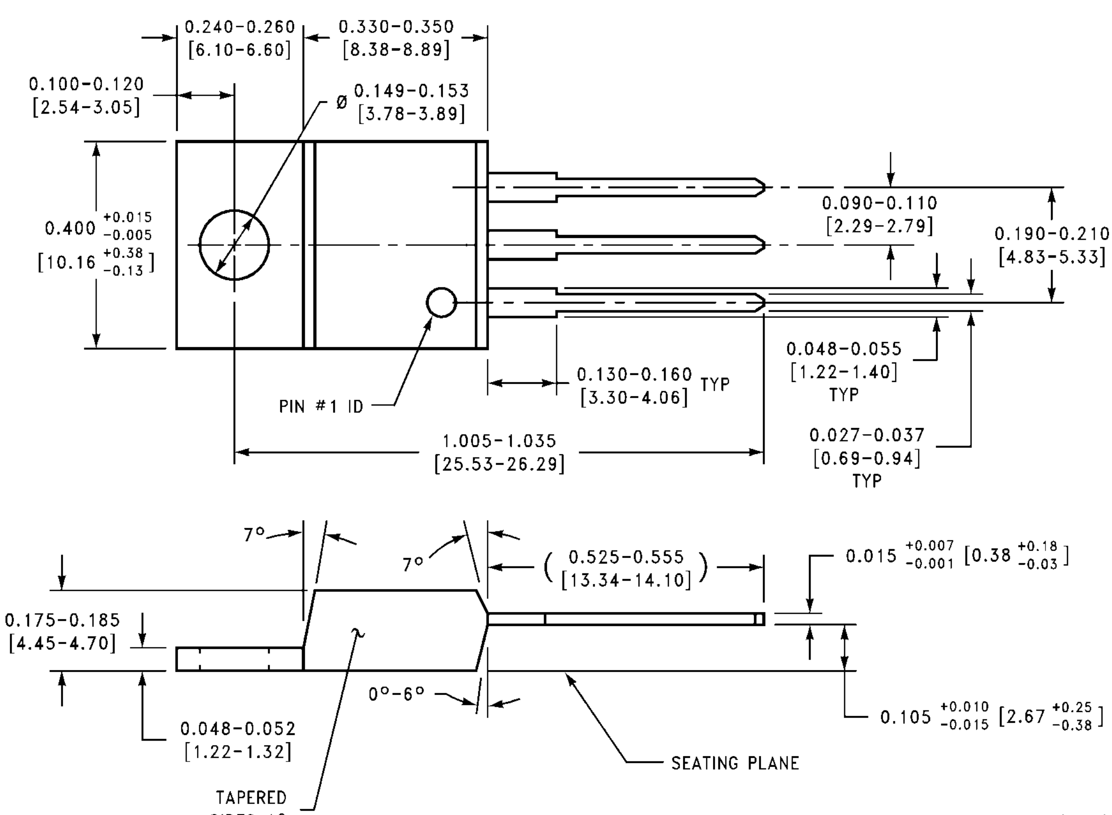
10094758

Remote Sensing

Physical Dimensions inches (millimeters) unless otherwise noted



3-Lead TO-263
NS Package Number TS3B



3-Lead TO-220
NS Package Number T03B

Notes

IMPORTANT NOTICE

Products		**Applications**	
Audio	www.ti.com/audio	Automotive and Transportation	www.ti.com/automotive
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